

MTB75N05HD

Preferred Device

Power MOSFET 75 Amps, 50 Volts N-Channel D²PAK

This Power MOSFET is designed to withstand high energy in the avalanche and commutation modes. The energy efficient design also offers a drain-to-source diode with a fast recovery time. Designed for low voltage, high speed switching applications in power supplies, converters and PWM motor controls, these devices are particularly well suited for bridge circuits where diode speed and commutating safe operating areas are critical and offer additional safety margin against unexpected voltage transients.

- Avalanche Energy Specified
- Source-to-Drain Diode Recovery Time Comparable to a Discrete Fast Recovery Diode
- Diode is Characterized for Use in Bridge Circuits
- I_{DSS} and $V_{DS(on)}$ Specified at Elevated Temperature
- Short Heatsink Tab Manufactured – Not Sheared
- Specially Designed Leadframe for Maximum Power Dissipation

MAXIMUM RATINGS ($T_C = 25^\circ\text{C}$ unless otherwise noted)

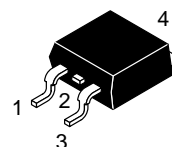
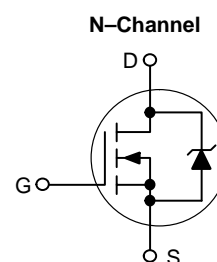
Rating	Symbol	Value	Unit
Drain-to-Source Voltage	V_{DSS}	50	Volts
Drain-to-Gate Voltage ($R_{GS} = 1.0\text{ M}\Omega$)	V_{DGR}	50	
Gate-to-Source Voltage – Continuous	V_{GS}	± 20	
Drain Current – Continuous – Continuous @ 100°C – Single Pulse ($t_p \leq 10\ \mu\text{s}$)	I_D	75	Amps
	I_D	65	
	I_{DM}	225	
Total Power Dissipation Derate above 25°C Total Power Dissipation @ $T_A = 25^\circ\text{C}$ (minimum footprint, FR-4 board)	P_D	125	Watts
		1.0	W/ $^\circ\text{C}$
		2.5	Watts
Operating and Storage Temperature Range	T_J, T_{stg}	-55 to 150	$^\circ\text{C}$
Single Pulse Drain-to-Source Avalanche Energy – Starting $T_J = 25^\circ\text{C}$ ($V_{DD} = 25\text{ V}$, $V_{GS} = 10\text{ V}$, Peak $I_L = 75\text{ A}$, $L = 0.177\text{ mH}$, $R_G = 25\ \Omega$)	E_{AS}	500	mJ
Thermal Resistance – Junction to Case – Junction to Ambient – Junction to Ambient (minimum footprint, FR-4 board)	$R_{\theta JC}$	1.0	$^\circ\text{C}/\text{W}$
	$R_{\theta JA}$	62.5	
	$R_{\theta JA}$	50	
Maximum Temperature for Soldering Purposes, 1/8" from case for 10 seconds	T_L	260	$^\circ\text{C}$



ON Semiconductor™

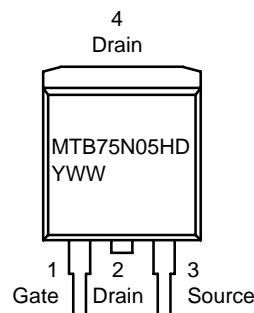
<http://onsemi.com>

75 AMPERES
50 VOLTS
 $R_{DS(on)} = 9.5\text{ m}\Omega$



D²PAK
CASE 418B
STYLE 2

MARKING DIAGRAM & PIN ASSIGNMENT



MTB75N05HD = Device Code
Y = Year
WW = Work Week

ORDERING INFORMATION

Device	Package	Shipping
MTB75N05HD	D ² PAK	50 Units/Rail
MTB75N05HDT4	D ² PAK	800/Tape & Reel

Preferred devices are recommended choices for future use and best overall value.

MTB75N05HD

ELECTRICAL CHARACTERISTICS (T_J = 25°C unless otherwise noted)

Characteristic	Symbol	Min	Typ	Max	Unit
----------------	--------	-----	-----	-----	------

OFF CHARACTERISTICS

Drain-to-Source Breakdown Voltage (V _{GS} = 0, I _D = 250 μAdc) Temperature Coefficient (Positive)	(C _{pk} ≥ 2) (Note 2.) V(BR)DSS	50 –	– 54.9	– –	Vdc mV/°C
Zero Gate Voltage Drain Current (V _{DS} = 50 V, V _{GS} = 0) (V _{DS} = 50 V, V _{GS} = 0, T _J = 125°C)	I _{DSS}	– –	– –	10 100	μAdc
Gate-Body Leakage Current (V _{GS} = ± 20 Vdc, V _{DS} = 0)	I _{GSS}	–	–	100	nAdc

ON CHARACTERISTICS (Note 1.)

Gate Threshold Voltage (V _{DS} = V _{GS} , I _D = 250 μAdc) Threshold Temperature Coefficient (Negative)	(C _{pk} ≥ 1.5) (Note 2.) V _{GS(th)}	2.0 –	– 6.3	4.0 –	Vdc mV/°C
Static Drain-to-Source On-Resistance (Note 3.) (V _{GS} = 10 Vdc, I _D = 20 Adc)	(C _{pk} ≥ 3.0) (Note 2.) R _{DS(on)}	–	7.0	9.5	mΩ
Drain-to-Source On-Voltage (V _{GS} = 10 Vdc) (Note 3.) (I _D = 75 A) (I _D = 20 Adc, T _J = 125°C)	V _{DS(on)}	– –	0.63 –	– 0.34	Vdc
Forward Transconductance (V _{DS} = 10 Vdc, I _D = 20 Adc)	g _{FS}	15	–	–	mhos

DYNAMIC CHARACTERISTICS (Note 2.)

Input Capacitance	(V _{DS} = 25 V, V _{GS} = 0, f = 1.0 MHz) (C _{pk} ≥ 2.0) (C _{pk} ≥ 2.0) (C _{pk} ≥ 2.0)	C _{iSS}	–	2600	3900	pF
Output Capacitance		C _{oSS}	–	1000	1300	
Transfer Capacitance		C _{rSS}	–	230	300	

SWITCHING CHARACTERISTICS (Note 4.)

Turn-On Delay Time	(V _{DD} = 25 V, I _D = 75 A, V _{GS} = 10 V, R _G = 9.1 Ω)	t _{d(on)}	–	15	30	ns
Rise Time		t _r	–	170	340	
Turn-Off Delay Time		t _{d(off)}	–	70	140	
Fall Time		t _f	–	100	200	
Gate Charge	(V _{DS} = 40 V, I _D = 75 A, V _{GS} = 10 V)	Q _T	–	71	100	nC
		Q ₁	–	13	–	
		Q ₂	–	33	–	
		Q ₃	–	26	–	

SOURCE-DRAIN DIODE CHARACTERISTICS

Forward On-Voltage (Note 2.)	(I _S = 75 A, V _{GS} = 0) (C _{pk} ≥ 10) (I _S = 20 A, V _{GS} = 0) (I _S = 20 A, V _{GS} = 0, T _J = 125°C)	V _{SD}	– – –	0.97 0.80 0.68	– 1.00 –	Vdc
Reverse Recovery Time	(I _S = 37.5 A, V _{GS} = 0, dI _S /dt = 100 A/μs)	t _{rr}	–	57	–	ns
		t _a	–	40	–	
		t _b	–	17	–	
Reverse Recovery Stored Charge		Q _{RR}	–	0.17	–	μC

INTERNAL PACKAGE INDUCTANCE

Internal Drain Inductance (Measured from contact screw on tab to center of die) (Measured from drain lead 0.25" from package to center of die)	L _D	– –	3.5 4.5	– –	nH
Internal Source Inductance (Measured from the source lead 0.25" from package to source bond pad)	L _S	–	7.5	–	

1. Pulse Test: Pulse Width ≤ 300 μs, Duty Cycle ≤ 2%.
2. Reflects Typical Values. C_{pk} = Absolute Value of (SPEC – AVG) / 3 * SIGMA).
3. For accurate measurements, good Kelvin contact required.
4. Switching characteristics are independent of operating junction temperature.

MTB75N05HD

TYPICAL ELECTRICAL CHARACTERISTICS (Note 5.)

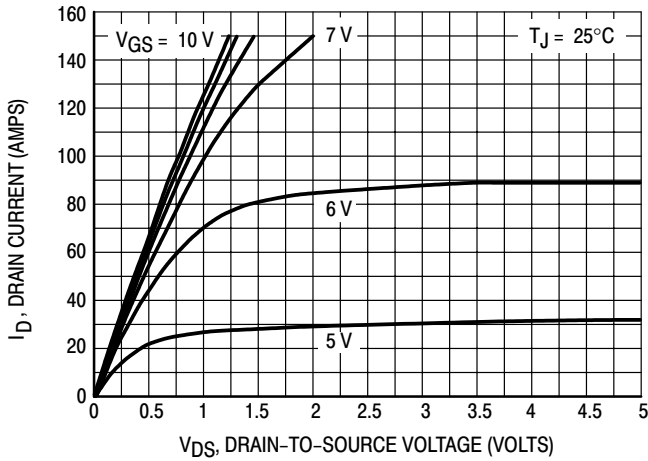


Figure 1. On-Region Characteristics

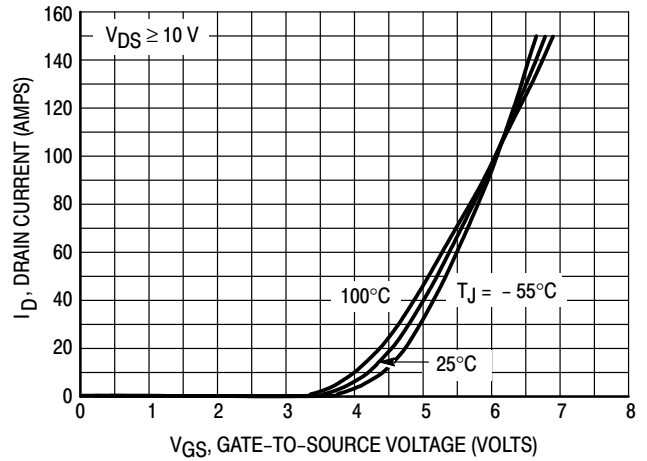


Figure 2. Transfer Characteristics

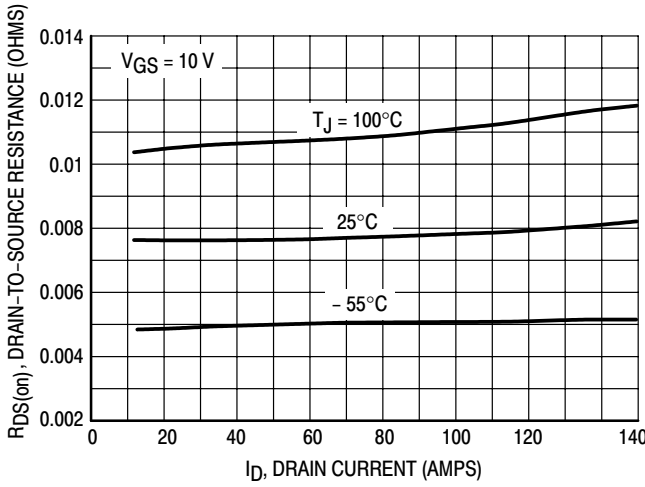


Figure 3. On-Resistance versus Drain Current and Temperature

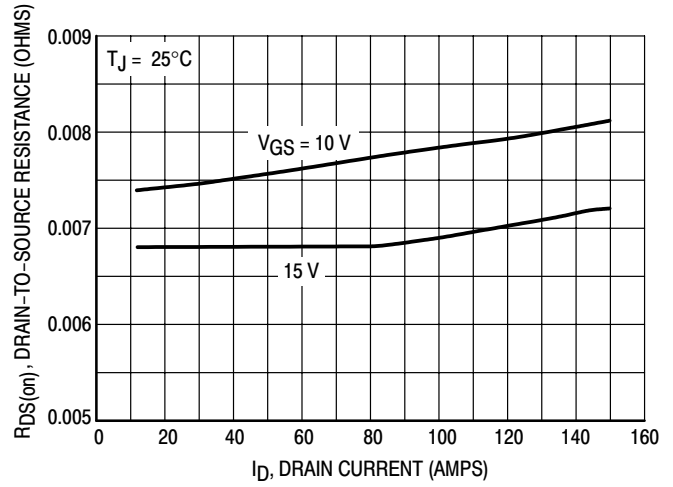


Figure 4. On-Resistance versus Drain Current and Gate Voltage

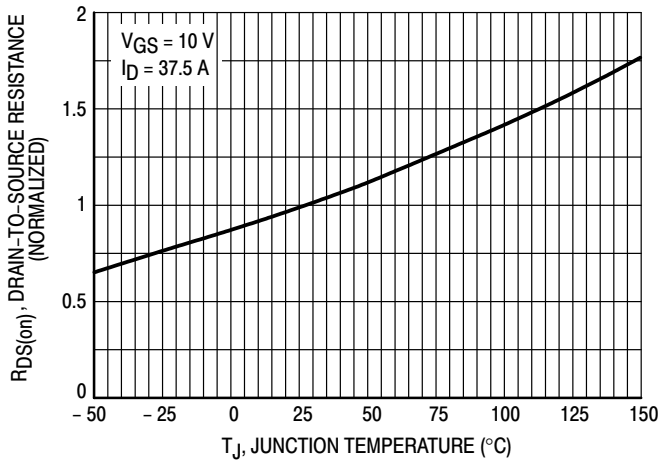


Figure 5. On-Resistance Variation with Temperature

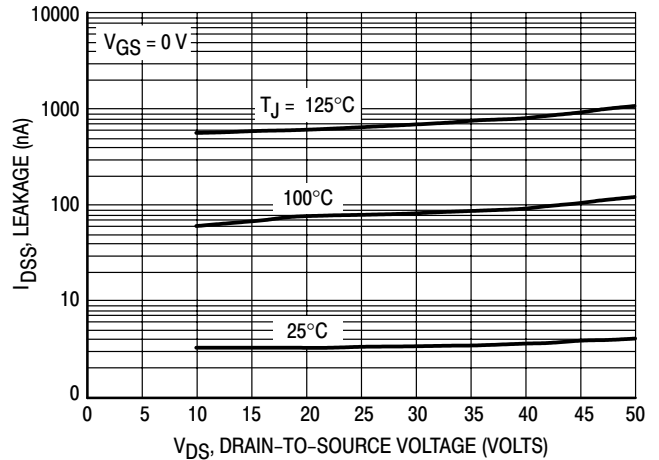


Figure 6. Drain-to-Source Leakage Current versus Voltage

5. Pulse Tests: Pulse Width $\leq 250 \mu\text{s}$, Duty Cycle $\leq 2\%$.

POWER MOSFET SWITCHING

Switching behavior is most easily modeled and predicted by recognizing that the power MOSFET is charge controlled. The lengths of various switching intervals (Δt) are determined by how fast the FET input capacitance can be charged by current from the generator.

The published capacitance data is difficult to use for calculating rise and fall because drain-gate capacitance varies greatly with applied voltage. Accordingly, gate charge data is used. In most cases, a satisfactory estimate of average input current ($I_{G(AV)}$) can be made from a rudimentary analysis of the drive circuit so that

$$t = Q/I_{G(AV)}$$

During the rise and fall time interval when switching a resistive load, V_{GS} remains virtually constant at a level known as the plateau voltage, V_{SGP} . Therefore, rise and fall times may be approximated by the following:

$$t_r = Q_2 \times R_G / (V_{GG} - V_{SGP})$$

$$t_f = Q_2 \times R_G / V_{SGP}$$

where

V_{GG} = the gate drive voltage, which varies from zero to V_{GG}

R_G = the gate drive resistance

and Q_2 and V_{SGP} are read from the gate charge curve.

During the turn-on and turn-off delay times, gate current is not constant. The simplest calculation uses appropriate values from the capacitance curves in a standard equation for voltage change in a RC network. The equations are:

$$t_{d(on)} = R_G C_{iss} \ln [V_{GG} / (V_{GG} - V_{SGP})]$$

$$t_{d(off)} = R_G C_{iss} \ln (V_{GG} / V_{SGP})$$

The capacitance (C_{iss}) is read from the capacitance curve at a voltage corresponding to the off-state condition when calculating $t_{d(on)}$ and is read at a voltage corresponding to the on-state when calculating $t_{d(off)}$.

At high switching speeds, parasitic circuit elements complicate the analysis. The inductance of the MOSFET source lead, inside the package and in the circuit wiring which is common to both the drain and gate current paths, produces a voltage at the source which reduces the gate drive current. The voltage is determined by $L di/dt$, but since di/dt is a function of drain current, the mathematical solution is complex. The MOSFET output capacitance also complicates the mathematics. And finally, MOSFETs have finite internal gate resistance which effectively adds to the resistance of the driving source, but the internal resistance is difficult to measure and, consequently, is not specified.

The resistive switching time variation versus gate resistance (Figure 9) shows how typical switching performance is affected by the parasitic circuit elements. If the parasitics were not present, the slope of the curves would maintain a value of unity regardless of the switching speed. The circuit used to obtain the data is constructed to minimize common inductance in the drain and gate circuit loops and is believed readily achievable with board-mounted components. Most power electronic loads are inductive; the data in the figure is taken with a resistive load, which approximates an optimally snubbed inductive load. Power MOSFETs may be safely operated into an inductive load; however, snubbing reduces switching losses.

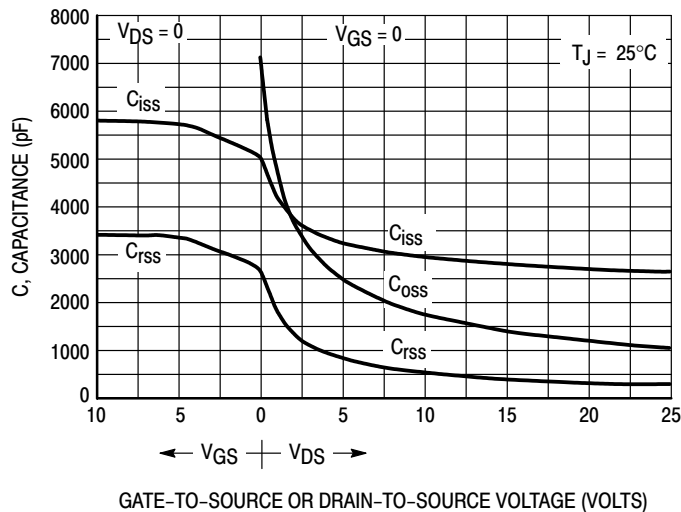


Figure 7. Capacitance Variation

MTB75N05HD

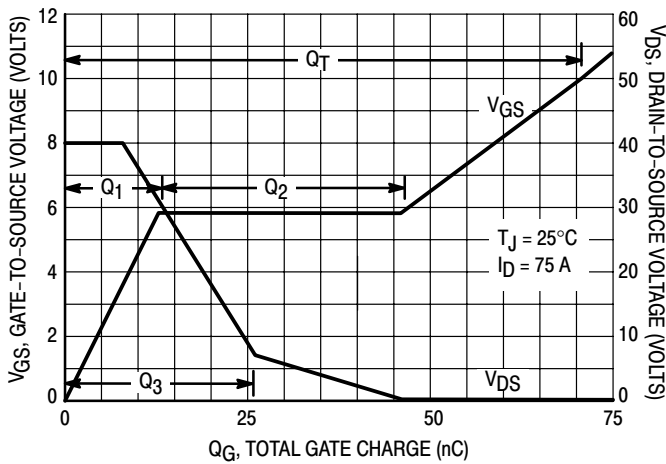


Figure 8. Gate-To-Source and Drain-To-Source Voltage versus Total Charge

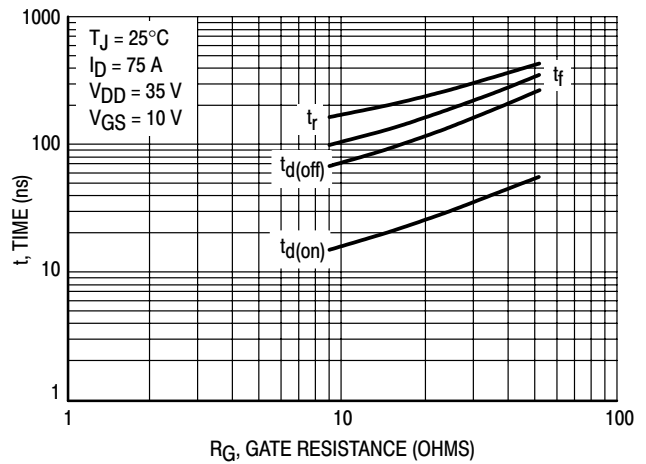


Figure 9. Resistive Switching Time Variation versus Gate Resistance

DRAIN-TO-SOURCE DIODE CHARACTERISTICS

The switching characteristics of a MOSFET body diode are very important in systems using it as a freewheeling or commutating diode. Of particular interest are the reverse recovery characteristics which play a major role in determining switching losses, radiated noise, EMI and RFI.

System switching losses are largely due to the nature of the body diode itself. The body diode is a minority carrier device, therefore it has a finite reverse recovery time, t_{rr} , due to the storage of minority carrier charge, Q_{RR} , as shown in the typical reverse recovery wave form of Figure 12. It is this stored charge that, when cleared from the diode, passes through a potential and defines an energy loss. Obviously, repeatedly forcing the diode through reverse recovery further increases switching losses. Therefore, one would like a diode with short t_{rr} and low Q_{RR} specifications to minimize these losses.

The abruptness of diode reverse recovery effects the amount of radiated noise, voltage spikes, and current ringing. The mechanisms at work are finite irremovable circuit parasitic inductances and capacitances acted upon by

high di/dt s. The diode's negative di/dt during t_a is directly controlled by the device clearing the stored charge. However, the positive di/dt during t_b is an uncontrollable diode characteristic and is usually the culprit that induces current ringing. Therefore, when comparing diodes, the ratio of t_b/t_a serves as a good indicator of recovery abruptness and thus gives a comparative estimate of probable noise generated. A ratio of 1 is considered ideal and values less than 0.5 are considered snappy.

Compared to ON Semiconductor standard cell density low voltage MOSFETs, high cell density MOSFET diodes are faster (shorter t_{rr}), have less stored charge and a softer reverse recovery characteristic. The softness advantage of the high cell density diode means they can be forced through reverse recovery at a higher di/dt than a standard cell MOSFET diode without increasing the current ringing or the noise generated. In addition, power dissipation incurred from switching the diode will be less due to the shorter recovery time and lower switching losses.

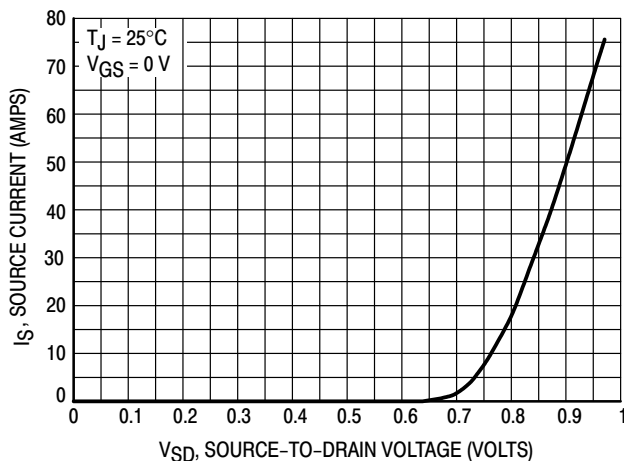


Figure 10. Diode Forward Voltage versus Current

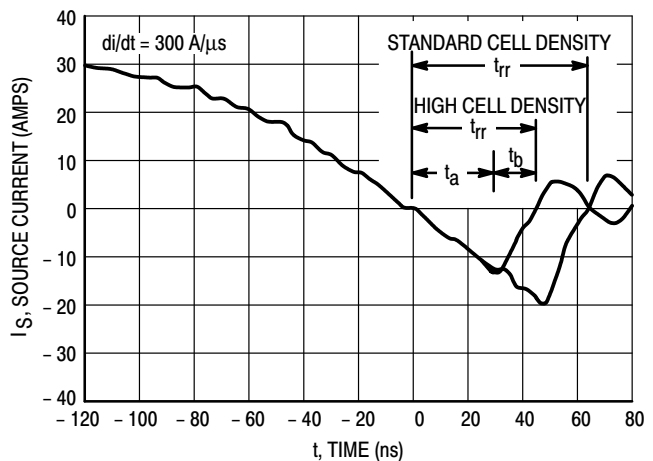


Figure 11. Reverse Recovery Time (t_{rr})

SAFE OPERATING AREA

The Forward Biased Safe Operating Area curves define the maximum simultaneous drain-to-source voltage and drain current that a transistor can handle safely when it is forward biased. Curves are based upon maximum peak junction temperature and a case temperature (T_C) of 25°C. Peak repetitive pulsed power limits are determined by using the thermal response data in conjunction with the procedures discussed in AN569, "Transient Thermal Resistance – General Data and Its Use."

Switching between the off-state and the on-state may traverse any load line provided neither rated peak current (I_{DM}) nor rated voltage (V_{DSS}) is exceeded, and that the transition time (t_r , t_f) does not exceed 10 μ s. In addition the total power averaged over a complete switching cycle must not exceed $(T_{J(MAX)} - T_C)/(R_{\theta JC})$.

A power MOSFET designated E-FET can be safely used in switching circuits with unclamped inductive loads. For

reliable operation, the stored energy from circuit inductance dissipated in the transistor while in avalanche must be less than the rated limit and must be adjusted for operating conditions differing from those specified. Although industry practice is to rate in terms of energy, avalanche energy capability is not a constant. The energy rating decreases non-linearly with an increase of peak current in avalanche and peak junction temperature.

Although many E-FETs can withstand the stress of drain-to-source avalanche at currents up to rated pulsed current (I_{DM}), the energy rating is specified at rated continuous current (I_D), in accordance with industry custom. The energy rating must be derated for temperature as shown in the accompanying graph (Figure 12). Maximum energy at currents below rated continuous I_D can safely be assumed to equal the values indicated.

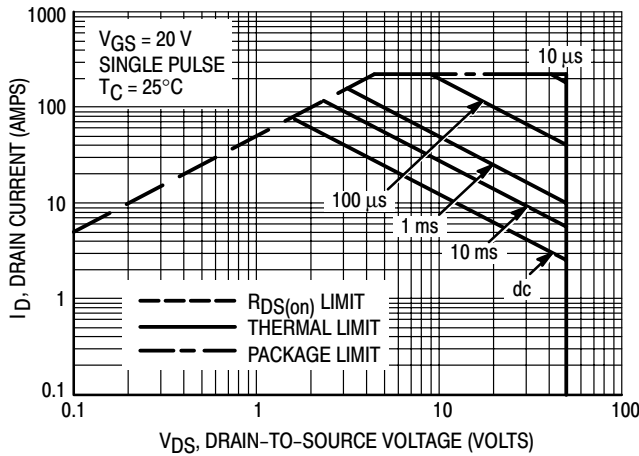


Figure 12. Maximum Rated Forward Biased Safe Operating Area

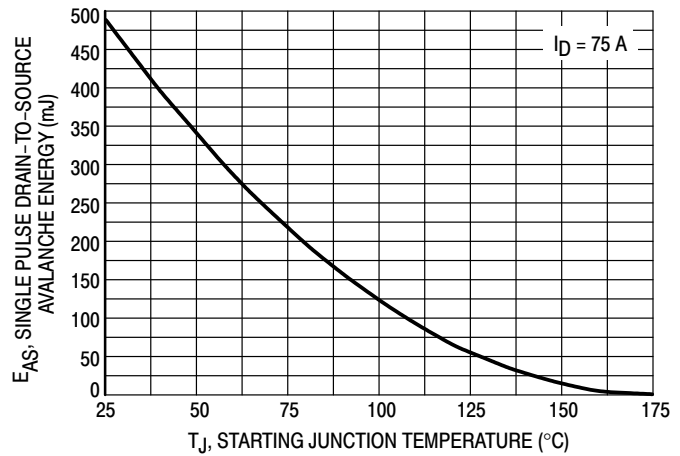


Figure 13. Maximum Avalanche Energy versus Starting Junction Temperature

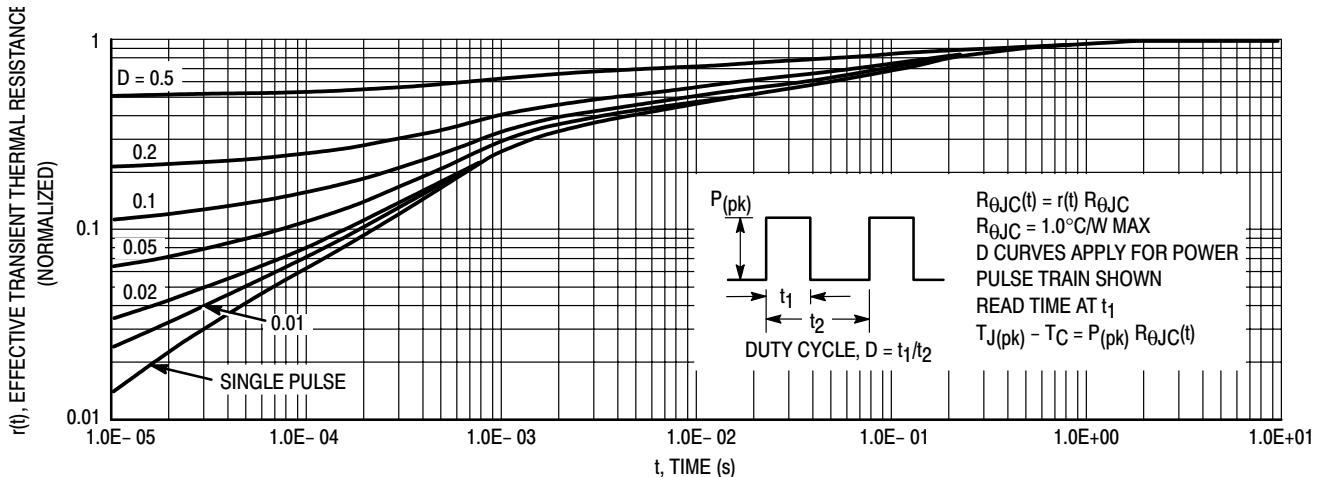


Figure 14. Thermal Response

MTB75N05HD

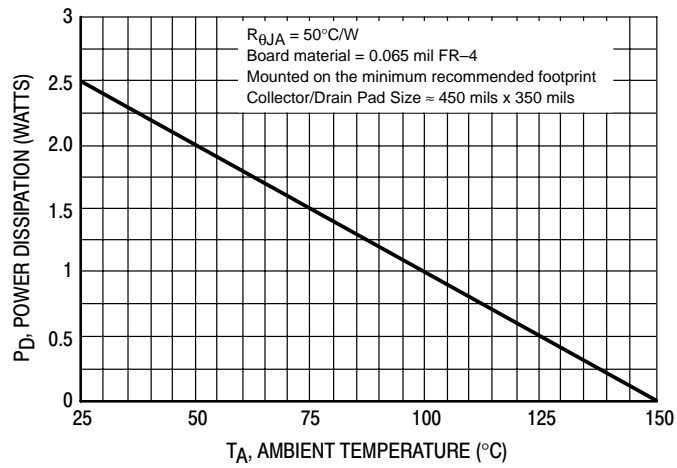
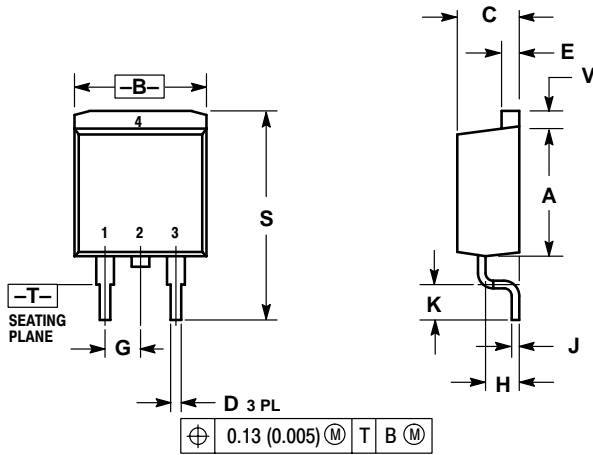


Figure 15. D²PAK Power Derating Curve

PACKAGE DIMENSIONS


D²PAK
CASE 418B-03
ISSUE D



- NOTES:
1. DIMENSIONING AND TOLERANCING PER ANSI Y14.5M, 1982.
2. CONTROLLING DIMENSION: INCH.

DIM	INCHES		MILLIMETERS	
	MIN	MAX	MIN	MAX
A	0.340	0.380	8.64	9.65
B	0.380	0.405	9.65	10.29
C	0.160	0.190	4.06	4.83
D	0.020	0.035	0.51	0.89
E	0.045	0.055	1.14	1.40
G	0.100 BSC		2.54 BSC	
H	0.080	0.110	2.03	2.79
J	0.018	0.025	0.46	0.64
K	0.090	0.110	2.29	2.79
S	0.575	0.625	14.60	15.88
V	0.045	0.055	1.14	1.40

- STYLE 2:
PIN 1. GATE
2. DRAIN
3. SOURCE
4. DRAIN

ON Semiconductor and  are trademarks of Semiconductor Components Industries, LLC (SCILLC). SCILLC reserves the right to make changes without further notice to any products herein. SCILLC makes no warranty, representation or guarantee regarding the suitability of its products for any particular purpose, nor does SCILLC assume any liability arising out of the application or use of any product or circuit, and specifically disclaims any and all liability, including without limitation special, consequential or incidental damages. "Typical" parameters which may be provided in SCILLC data sheets and/or specifications can and do vary in different applications and actual performance may vary over time. All operating parameters, including "Typicals" must be validated for each customer application by customer's technical experts. SCILLC does not convey any license under its patent rights nor the rights of others. SCILLC products are not designed, intended, or authorized for use as components in systems intended for surgical implant into the body, or other applications intended to support or sustain life, or for any other application in which the failure of the SCILLC product could create a situation where personal injury or death may occur. Should Buyer purchase or use SCILLC products for any such unintended or unauthorized application, Buyer shall indemnify and hold SCILLC and its officers, employees, subsidiaries, affiliates, and distributors harmless against all claims, costs, damages, and expenses, and reasonable attorney fees arising out of, directly or indirectly, any claim of personal injury or death associated with such unintended or unauthorized use, even if such claim alleges that SCILLC was negligent regarding the design or manufacture of the part. SCILLC is an Equal Opportunity/Affirmative Action Employer.

PUBLICATION ORDERING INFORMATION

NORTH AMERICA Literature Fulfillment:

Literature Distribution Center for ON Semiconductor
P.O. Box 5163, Denver, Colorado 80217 USA
Phone: 303-675-2175 or 800-344-3860 Toll Free USA/Canada
Fax: 303-675-2176 or 800-344-3867 Toll Free USA/Canada
Email: ONlit@hibbertco.com
Fax Response Line: 303-675-2167 or 800-344-3810 Toll Free USA/Canada

N. American Technical Support: 800-282-9855 Toll Free USA/Canada

EUROPE: LDC for ON Semiconductor – European Support

German Phone: (+1) 303-308-7140 (Mon-Fri 2:30pm to 7:00pm CET)
Email: ONlit-german@hibbertco.com
French Phone: (+1) 303-308-7141 (Mon-Fri 2:00pm to 7:00pm CET)
Email: ONlit-french@hibbertco.com
English Phone: (+1) 303-308-7142 (Mon-Fri 12:00pm to 5:00pm GMT)
Email: ONlit@hibbertco.com

EUROPEAN TOLL-FREE ACCESS*: 00-800-4422-3781

*Available from Germany, France, Italy, UK, Ireland

CENTRAL/SOUTH AMERICA:

Spanish Phone: 303-308-7143 (Mon-Fri 8:00am to 5:00pm MST)
Email: ONlit-spanish@hibbertco.com
Toll-Free from Mexico: Dial 01-800-288-2872 for Access –
then Dial 866-297-9322

ASIA/PACIFIC: LDC for ON Semiconductor – Asia Support

Phone: 303-675-2121 (Tue-Fri 9:00am to 1:00pm, Hong Kong Time)
Toll Free from Hong Kong & Singapore:
001-800-4422-3781

Email: ONlit-asia@hibbertco.com

JAPAN: ON Semiconductor, Japan Customer Focus Center

4-32-1 Nishi-Gotanda, Shinagawa-ku, Tokyo, Japan 141-0031
Phone: 81-3-5740-2700
Email: r14525@onsemi.com

ON Semiconductor Website: <http://onsemi.com>

For additional information, please contact your local Sales Representative.