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**Quad SPST, CMOS RF/Video Switch**

The IH5352 is a quad SPST, CMOS monolithic switch which uses a "Series/Shunt" ("T" switch) configuration to obtain high OFF isolation while maintaining good frequency response in the ON condition.

Construction of remote and portable video equipment with extended battery life is facilitated by the extremely low current requirements. Switching speeds are typically  $t_{ON} = 150ns$  and  $t_{OFF} = 80ns$ . "Break-Before-Make" switching is guaranteed.

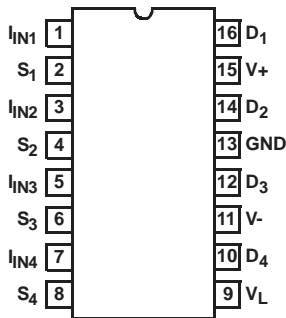
Switch ON resistance is typically  $40\Omega - 50\Omega$  with  $\pm 15V$  power supplies, increasing to typically  $175\Omega$  for  $\pm 5V$  supplies.

**Part Number Information**

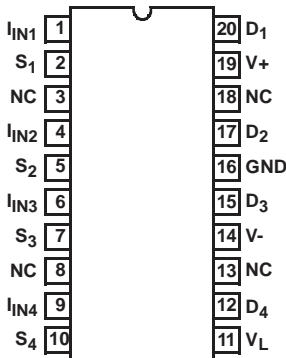
PART NUMBER	TEMP. RANGE (°C)	PACKAGE	PKG. NO.
IH5352CPE	0 to 70	16 Ld PDIP	E16.3
IH5352CBP	0 to 70	20 Ld SOIC	M20.3

**Pinouts**

**IH5352 (PDIP) TOP VIEW**



**IH5352 (SOIC) TOP VIEW**



**Features**

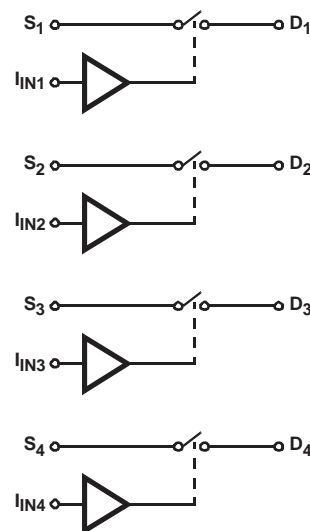
- $r_{DS(ON)}$  .....  $75\Omega$
- Switch Attenuation Varies Less Than 3dB From DC to 100MHz
- OFF Isolation at 10MHz .....  $>70dB$
- Crosstalk Isolation at 10MHz .....  $>60dB$
- Compatible With TTL, CMOS Logic
- Wide Operating Power Supply Range
- Power Supply Current .....  $<1\mu A$
- "Break-Before-Make" Switching
- Fast Switching (Typ) .....  $80ns/150ns$

**Applications**

- Video Switch
- Communications Equipment
- Disk Drives
- Instrumentation
- CATV

**Functional Diagram**

SWITCH STATE SHOWN FOR LOGIC "0" INPUT

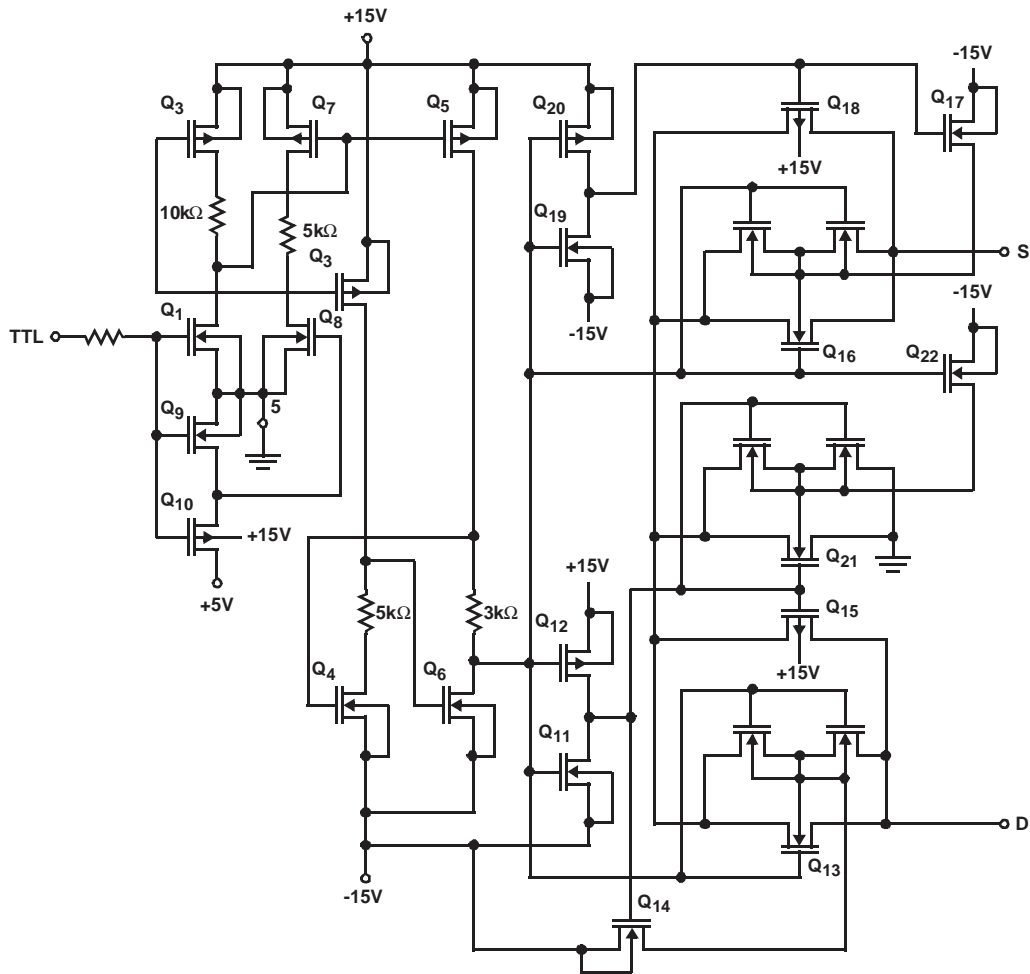


TRUTH TABLE

LOGIC	SWITCHES
0	Off
1	On

Schematic Diagram

1/4 IH5352



**Absolute Maximum Ratings**

V+ to Ground	+18V
V- to Ground	-18V
V <sub>L</sub> to Ground	V+ to V-
Logic Control Voltage	V+ to V-
Analog Input Voltage	V+ to V-
Current (Any Terminal)	50mA

**Thermal Information**

Thermal Resistance (Typical, Note 1)	$\theta_{JA}$ (°C/W)
SOIC Package	90
PDIP Package	90
Maximum Junction Temperature (Plastic Packages)	150°C
Maximum Storage Temperature Range	-65°C to 150°C
Maximum Lead Temperature (Soldering, 10s)	300°C (SOIC - Lead Tips Only)

**Operating Conditions**

Temperature Range	0°C to 70°C
Supply Voltage Range	
V+, V <sub>L</sub>	5V to 15V
V-	-5V to -15V

CAUTION: Stresses above those listed in "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress only rating and operation of the device at these or any other conditions above those indicated in the operational sections of this specification is not implied.

NOTE:

1.  $\theta_{JA}$  is measured with the component mounted on an evaluation PC board in free air.

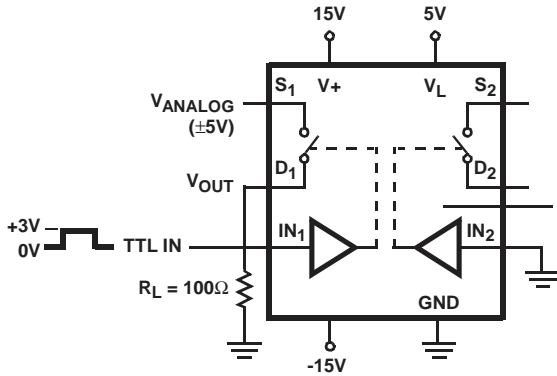
**Electrical Specifications** V+ = +15V, V<sub>L</sub> = +5V, V- = -15V, Unless Otherwise Specified

PARAMETER	TEST CONDITIONS	(NOTE 2) TYP 25°C	(NOTE 4)			UNITS
			0°C	25°C	70°C	
<b>DYNAMIC CHARACTERISTICS</b>						
Turn ON Time, t <sub>ON</sub>	Figure 1	150	-	-	-	ns
Turn OFF Time, t <sub>OFF</sub>		80	-	-	-	ns
OFF Isolation, OIRR	Figure 2	70	-	-	-	dB
Crosstalk, CCRR	Figure 3	-60	-	-	-	dB
Switch Attenuation 3dB Frequency, f <sub>3dB</sub>	Figure 4	100	-	-	-	MHz
<b>DIGITAL INPUT CHARACTERISTICS</b>						
Logic "1" Input Voltage, V <sub>IH</sub>		>2.4	-	-	-	V
Logic "0" Input Voltage, V <sub>IL</sub>		<0.8	-	-	-	V
Input Logic Current, I <sub>IN</sub>	V <sub>IN</sub> > 2.4V or < 0V	0.1	±1	±1	10	µA
<b>ANALOG SWITCH CHARACTERISTICS</b>						
Drain-Source ON Resistance, r <sub>DS(ON)</sub>	V <sub>D</sub> = ±5V, I <sub>S</sub> = 10mA, V <sub>IN</sub> ≥ 2.4V	50	75	75	100	Ω
	V <sub>D</sub> = ±10V, I <sub>S</sub> = 10mA, V <sub>IN</sub> ≥ 2.4V	100	150	150	175	Ω
	V+ = V <sub>L</sub> = +5V, V <sub>IN</sub> = 3V, V- = -5V, V <sub>D</sub> = ±3V, I <sub>S</sub> = 10mA	175	300	300	350	Ω
On Resistance Match Between Channels, Δr <sub>DS(ON)</sub>	I <sub>S</sub> = 10mA, V <sub>D</sub> = ±5V	5	-	-	-	Ω
Switch OFF Leakage Current, I <sub>D(OFF)</sub> or I <sub>S(OFF)</sub>	V <sub>S/D</sub> = ±5V or ±14V, V <sub>IN</sub> ≤ 0.8V (Note 3)	-	-	±2	100	nA
Switch ON Leakage Current, I <sub>D(ON)</sub> + I <sub>S(ON)</sub>	V <sub>S/D</sub> = ±5V or ±14V, V <sub>IN</sub> ≥ 2.4V	-	-	±2	100	nA
<b>POWER SUPPLY CHARACTERISTICS</b>						
Positive Supply Quiescent Current, I+	V <sub>IN</sub> = 0V or +5V	0.1	1	1	10	µA
Negative Supply Quiescent Current, I-		0.1	1	1	10	µA
Logic Supply Quiescent Current, I <sub>L</sub>		0.1	1	1	10	µA

NOTES:

2. Typical values are not tested in production. They are given as a design aid only.
3. Positive and negative voltages applied to opposite sides of switch, in both directions successively.
4. Min or Max value, unless otherwise specified.

Test Circuits and Waveforms



NOTE: Only one channel shown. Others act identically.

FIGURE 1A. TEST CIRCUIT

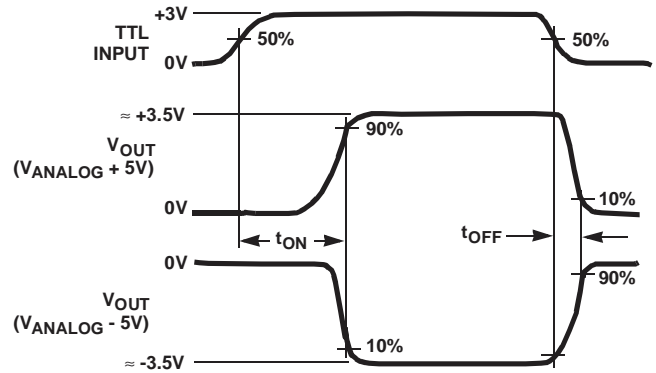
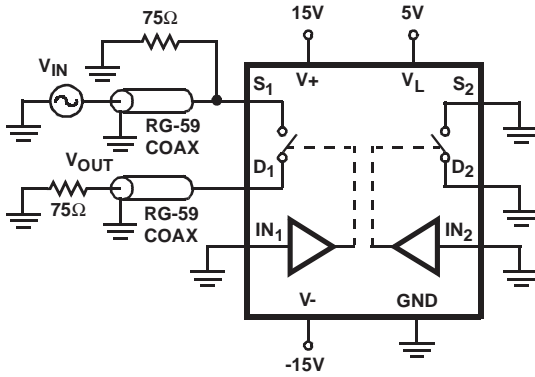


FIGURE 1B. MEASUREMENT POINTS

FIGURE 1. tON AND tOFF

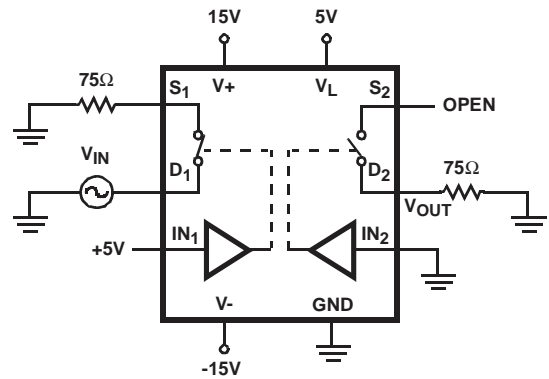


$$V_{IN} = \pm 5V(10V_{P-P}) \text{ at } f = 10\text{MHz}$$

$$OIRR = 20\text{Log} \frac{V_{IN}}{V_{OUT}}$$

NOTE: Only one channel shown. Others act identically.

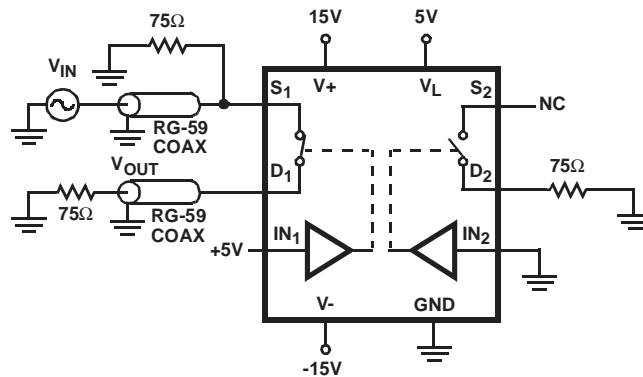
FIGURE 2. OFF ISOLATION TEST CIRCUIT



$$V_{IN} = 225\text{mV}_{RMS} \text{ at } f = 10\text{MHz}$$

$$CCRR = 20\text{Log} \frac{V_{OUT}}{V_{IN}}$$

FIGURE 3. CROSTALK TEST CIRCUIT



$$ATTN = 20\text{Log} \frac{R_L}{r_{DS(ON)} + R_L}$$

Nominally, at DC, ATTN equals -4dB. When the attenuation reaches -1dB, the frequency at which this occurs is f3dB.

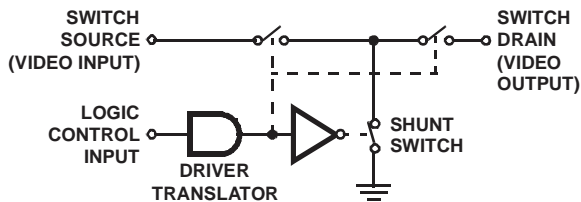
NOTE: Only one channel shown. Others act identically.

FIGURE 4. SWITCH ATTENUATION TEST CIRCUIT

## Detailed Description

Figure 5 shows the internal circuit of one channel of the IH5352. Here, a shunt switch is closed, and the two series switches are open when the video switch channel is open or off. This provides much better isolation between the input and output terminals than a simple series switch does, especially at high frequencies. The result is excellent off-isolation in the Video and RF frequency ranges when compared to conventional analog switches.

The control input level shifting circuitry is very similar to that of the IH5140 series of Analog Switches, and gives very high speed, guaranteed "Break-Before-Make" action, low static power consumption and TTL compatibility.



NOTE: 1 channel of 4 shown.

FIGURE 5. INTERNAL SWITCH CONFIGURATION

## Typical Applications

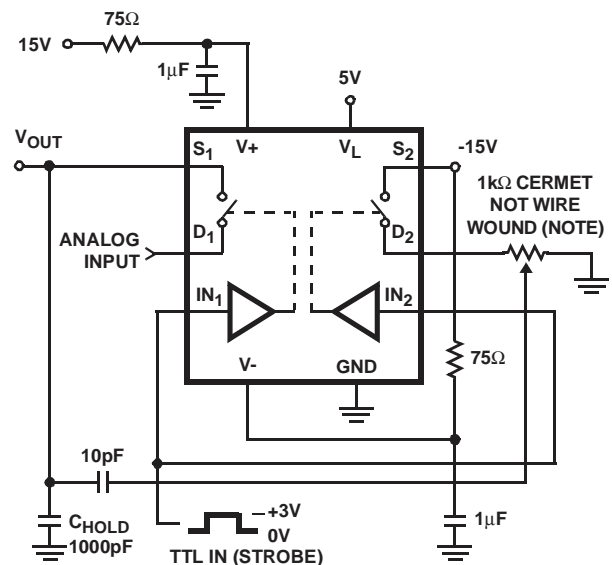
### Charge Compensation Techniques

Charge injection results from the signals out of the level translation circuit being coupled through the gate-channel and gate-source/drain capacitances to the switch inputs and outputs. This feedthrough is particularly troublesome in Sample-and-Hold or Track-and-Hold applications, as it causes a Sample (Track) to Hold offset. The IH5352 has a typical injected charge of 30pC-50pC (corresponding to 30mV-50mV on a 1000pF capacitor), at  $V_{S-D}$  of about 0V.

This Sample (Track) to Hold offset can be compensated by bringing in a signal equal in magnitude but of the opposite polarity. The circuit of Figure 6 accomplishes this charge injection compensation by using one side of the device as a S & H (T & H) switch, and the other side as a generator of a compensating signal. The 1k $\Omega$  potentiometer allows the user to adjust the net injected charge to exactly zero for any analog voltage in the -5V to +5V range.

Since individual parts are very consistent in their charge injection, it is possible to replace the potentiometer with a pair of fixed resistors, and achieve less than 5mV error for all devices without adjustment.

An alternative arrangement, using a standard TTL inverter to generate the required inversion, is shown in Figure 7. The capacitor needs to be increased, and becomes the only method of adjustment. A fixed value of 22pF is good for analog values referred to ground, while 35pF is optimum for AC coupled signals referred to -5V as shown in the figure. The choice of -5V is based on the virtual disappearance at this analog level of the transient component of switching charge injection. This combination will lead to a virtually "glitch-free" switch.



NOTE: Adjust pot for 0mV<sub>P-P</sub> step at  $V_{OUT}$  with no analog (AC) signal present.

FIGURE 6. CHARGE INJECTION COMPENSATION

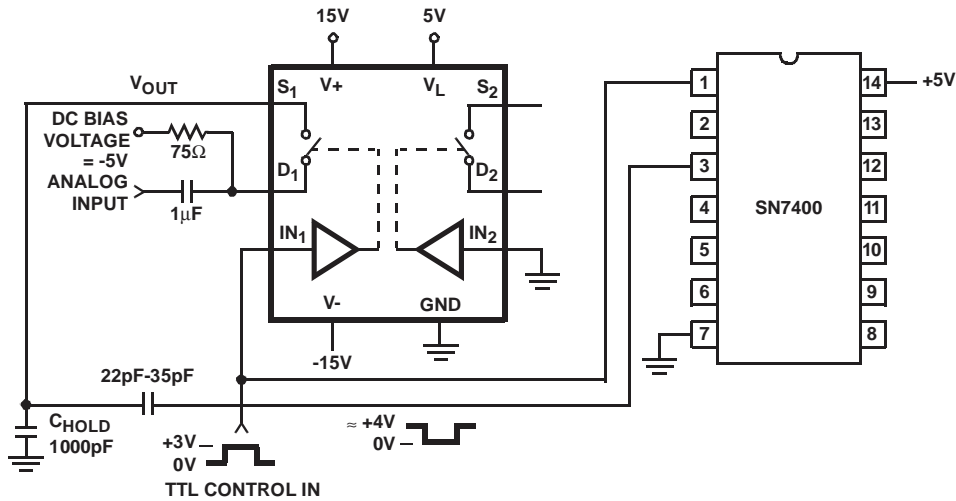


FIGURE 7. ALTERNATIVE COMPENSATION CIRCUIT

**Overvoltage Protection**

If sustained operation with no supplies but with analog signals applied is possible, it is recommended that diodes (such as 1N914) be inserted in series with the supply lines to the IH5352. Such conditions can occur if these signals come from a separate power supply or another location, for example. The diodes will be reverse biased under this type of operation, preventing heavy currents from flowing from the analog source through the IH5352.

The same method of protection will provide over  $\pm 25V$  overvoltage protection on the analog inputs when the supplies are present. The schematic for this connection is shown in Figure 8.

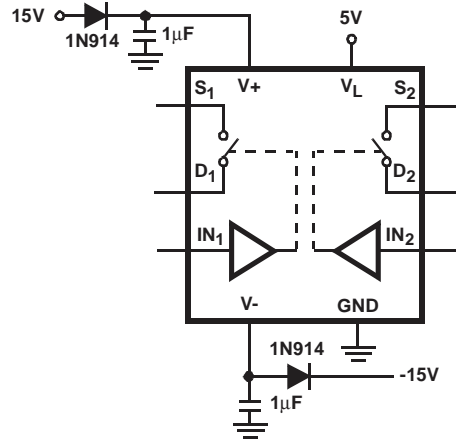


FIGURE 8. OVERVOLTAGE PROTECTION

Typical Performance Curves

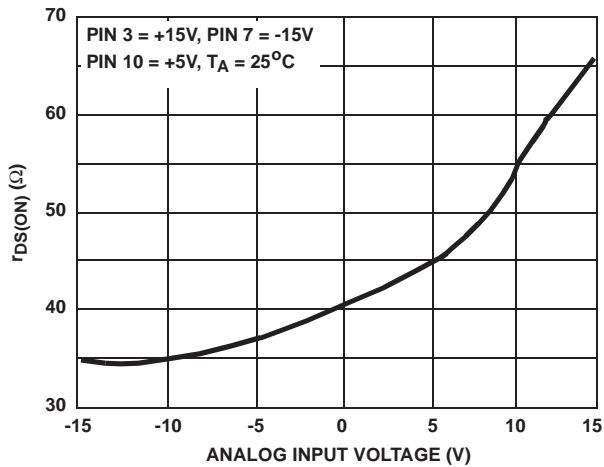


FIGURE 9.  $r_{DS(ON)}$  vs ANALOG INPUT VOLTAGE WITH  $\pm 15V$  POWER SUPPLIES

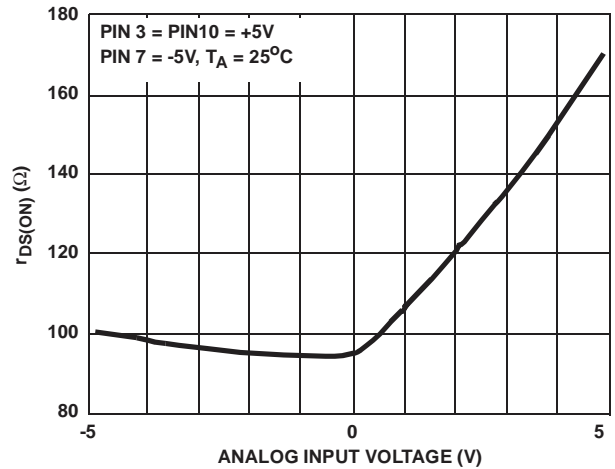


FIGURE 10.  $r_{DS(ON)}$  vs ANALOG INPUT LEVEL WITH  $\pm 5V$  POWER SUPPLIES

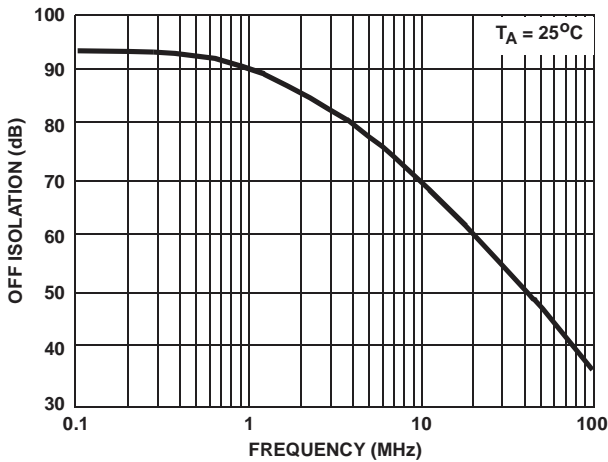


FIGURE 11. OFF ISOLATION vs FREQUENCY (SEE FIGURE 2)

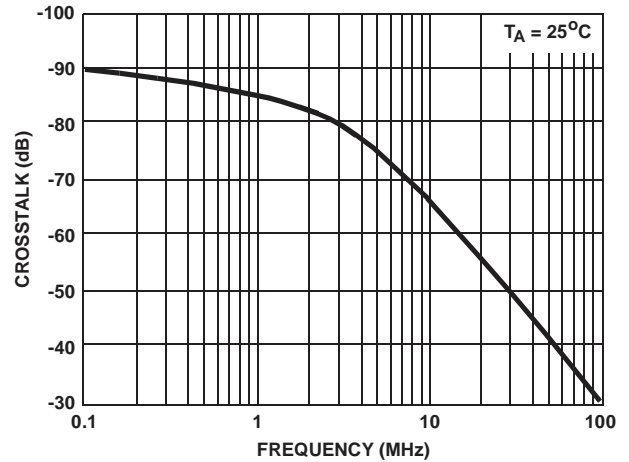


FIGURE 12. CROSSTALK vs FREQUENCY (SEE FIGURE 3)

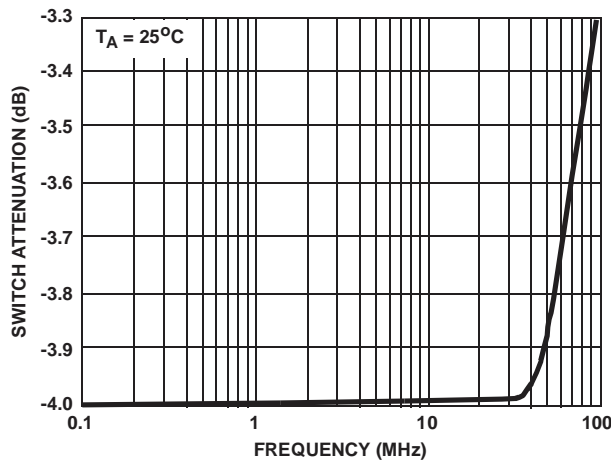


FIGURE 13. SWITCH ATTENUATION vs FREQUENCY ( $R_L = 75\Omega$ , SEE FIGURE 4)

**Die Characteristics**

**DIE DIMENSIONS:**

2617μm x 5233μm

**METALLIZATION:**

Type: Al  
 Thickness: 10kÅ ±1kÅ

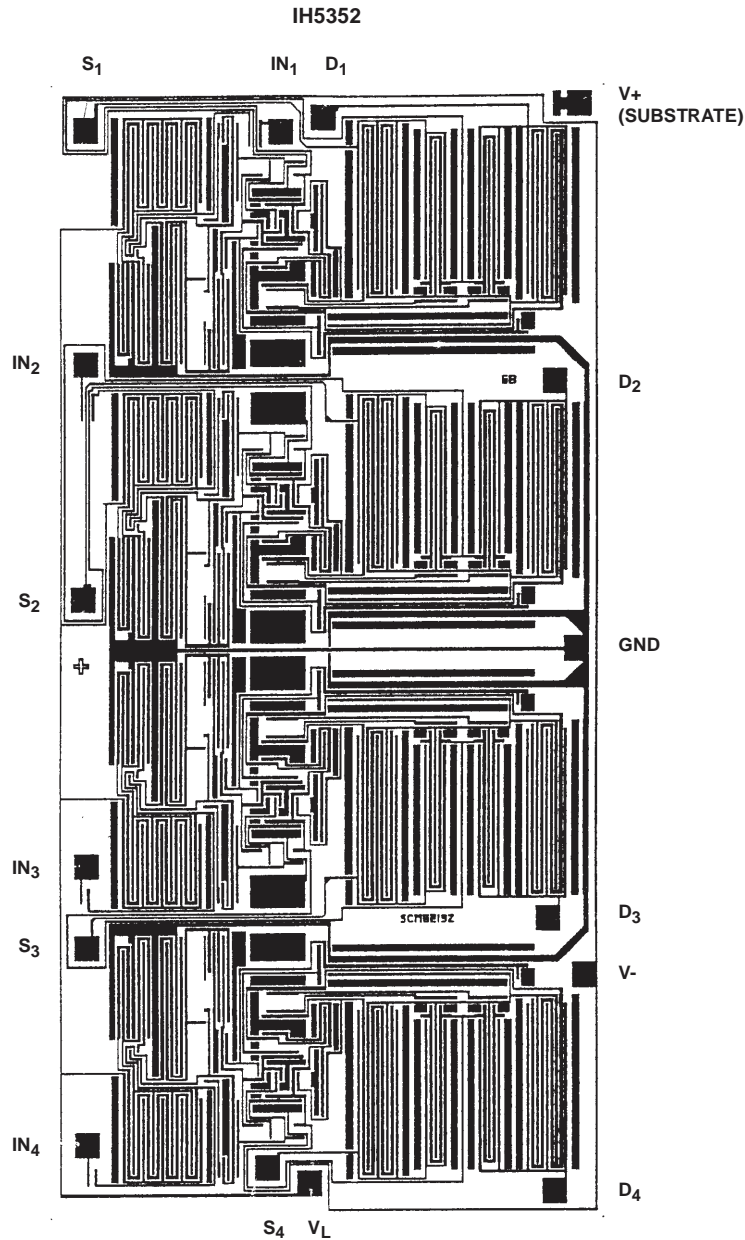
**PASSIVATION:**

Type: PSG/Nitride  
 PSG Thickness: 7kÅ ±1.4kÅ  
 Nitride Thickness: 8kÅ ±1.2kÅ

**WORST CASE CURRENT DENSITY:**

9.1 x 10<sup>4</sup> A/cm<sup>2</sup>

**Metallization Mask Layout**



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